

#### **Features**

- 2.5-5.5V Operation Voltage
- 0.1µA Ultra-low Current Shutdown Mode
- Improved Pop & Click Circuitry
- Unique Modulation Scheme Reduces EMI Emissions
- No Output Filter Required
- External Gain Configuration Capability
- 0.4V Compatible Threshold for /SHDN Pin
- BTL Output Supporting Capacitive Loads
- RoHS Compliant and 100% Lead(Pb)-Free
- Package: MSOP-8

### **Applications**

- PMP
- PSP
- Game Machine
- GPS Tracker
- Portable Electronic Devices

### **General Description**

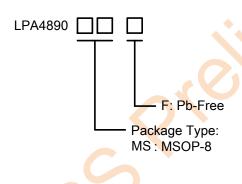
The LPA4890 is an audio power amplifier designed for portable communication device applications such as GPS trackers. The LPA4890 is capable of delivering 1.2W of continuous average power to an  $8\Omega$  load from a 5.2V power supply with less than 1% distortion (THD+N).

The LPA4890 is designed specifically to provide high quality output audio power with a minimal number of external components. The LPA4890 does not require output coupling capacitors or bootstrap capacitors, and therefore is ideally suited where minimal board or solution size is a primary requirement. It features a low-power shutdown mode, which is achieved by driving the /SHDN pin with logic low. Additionally, the LPA4890 features an internal thermal shutdown protection.

The LPA4890 contains circuitry to prevent "pop and click" noise that would occur during turn-on and turn-off transitions. For maximum flexibility, the LPA4890 provides an externally controlled gain (with resistors), as well as an externally controlled turn-on and turn-off times (with the bypass capacitor).

The LPA4890 is available in a MSOP-8 package.

## **Marking Information**



## **Ordering and Package Information**

Part Number	Top Mark	Package	T&R	
LPA4890MSF 4890M MSOP-		MSOP-8	3K/REEL	
Marking indication: Y: Production Year, W: Production week, X: Series Number				



# **Typical Application Circuits**

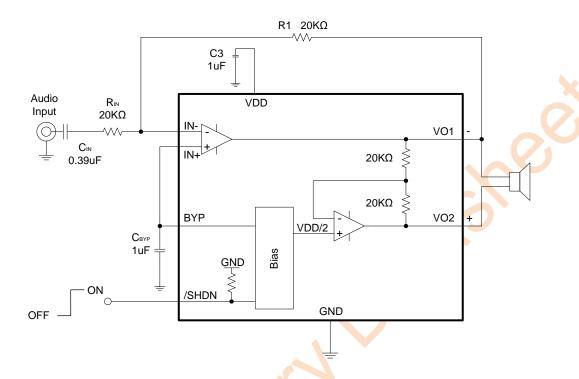


Figure 1-1. Typical Application Circuits with Single Input

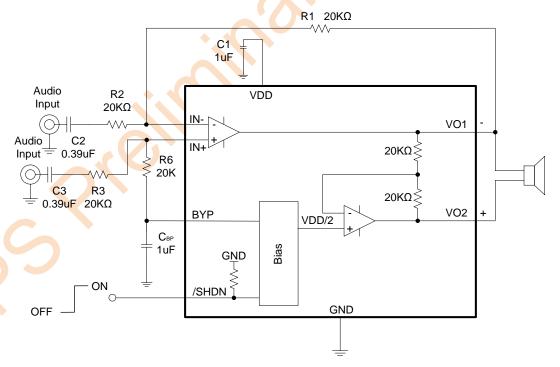
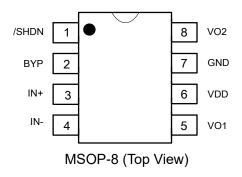


Figure 1-2. Typical Application Circuits with Differential Input





# **Pin Configuration**



# **Pin Description**

Pin	Name	Description		
1	/SHDN	Shutdown input pin. The device is enabled with this pin pulled high. The device enters the shutdown mode if this pin is floating or pulled low. This pin is internally pulled down to GND through the RPD resistor (refer to Figure 2 and Electrical Characteristics Table).		
2	BYP	Bypass capacitor pin. This pin provides the common mode voltage. Connect 1µF ceramic capacitor as close as possible to this pin.		
3	IN+	Positive input of the first amplifier. This pin receives the common mode voltage.		
4	IN-	Negative input of the first amplifier. This pin receives the audio input signal. Connect this pin to the feedback resistor R <sub>F</sub> and to the input resistor R <sub>IN</sub> (refer to Figure 2).		
5	VO1	Negative output. Connected this pin to the load and to the feedback resistor R <sub>F</sub> .		
6	VDD	VDD power supply input. Connect at least 1µF ceramic capacitor as close as possible to this pin.		
7	GND	Ground.		
8	VO2	Positive output.		



### **Functional Block Diagram**

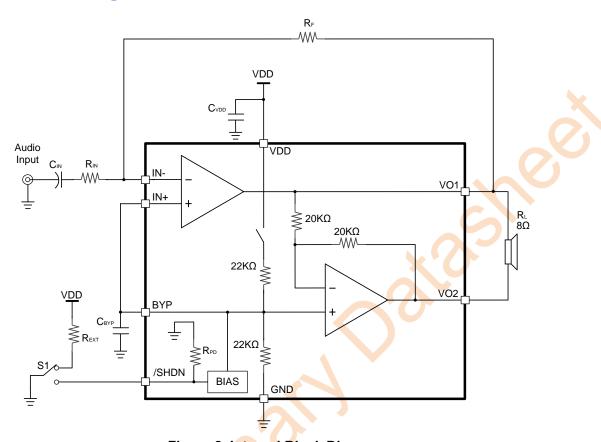


Figure 2. Internal Block Diagram

### Absolute Maximum Ratings (Note 1)

•	All pins to GND	0.3V to + 6V
•	Maximum Junction Temperature (TJMAX)	150°C
•	Storage Temperature Range, Tstg	65°C to 150°C
•	Maximum Soldering Temperature (at leads, 10 seconds)	260°C

\*Note 1: Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, instead of functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### Thermal Information

•	Maximum Power Dissipation ( $T_A \le 25^{\circ}C$ )		0.9W
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<sup>•</sup> Thermal Resistance (θJ<sub>A</sub>) (Note 2)------ 165°C/W

## **Recommended Operating Conditions**

	Input Voltage		2.5V to 5.5V
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Ambient Temperature ------ -20°C to 80°C

<sup>\*</sup>Note 2: It is based on 2S2P JEDEC standard PCB.





## **Electrical Characteristics**

The following parameters are guaranteed under condition  $V_{DD}$  =5V,  $T_A$  = -40°C to 85°C unless otherwise noted.  $T_A$  = 25°C for typical value.

Symbol	Parameter	Test Conditions		Min	Тур	Max	Unit
Vos	Output offset voltage (Measured differentially)	V <sub>i</sub> =0V, A <sub>v</sub> =2V/V, V <sub>DD</sub> =2.5V to 5.5V			3	20	mV
		V <sub>DD</sub> =5.5V, no load			4.4		
lα	Quiescent current	V <sub>DD</sub> =3.6V, no load			3.4	0	mA
		V <sub>DD</sub> =2.5V, no load		C	2.9		
Ishdn	Shutdown Current	V <sub>SHDN</sub> =0.35V, V <sub>DD</sub> =2.5V to 5.5V	1		0.1		μA
		V <sub>DD</sub> =5.5V, no load		400			
RDS(ON)	Static drain-source on-state resistance	V <sub>DD</sub> =3.6V, no load			500		mΩ
		V <sub>DD</sub> =2.5V, no load			700		-
Po	Output power	V <sub>DD</sub> =5.2V, R <sub>L</sub> =8Ω, THD=1%, f=1KHz			1.2		W
V <sub>SD_H</sub>	/SHDN pin logic High	V <sub>DD</sub> =2.5V to 5.5V		1.4			.,
V <sub>SD_L</sub>	/SHDN pin logic Low	V <sub>DD</sub> =2.5V to 5.5V				0.4	V
THD+N		V <sub>DD</sub> =5V, Po=1W, R <sub>L</sub> =8Ω, f=1KHz			0.3		
	Total harmonic distortion plus noise	V <sub>DD</sub> =3.6V, Po=0.5W, R <sub>L</sub> =8Ω, f=1KHz			0.4		%
		$V_{DD}$ =2.5V, Po=0.2W, R <sub>L</sub> =8 $\Omega$ , f=1KHz			0.45		
PSRR	Supply ripple rejection ratio	V <sub>DD</sub> =5V, inputs ac-grounded	f=217Hz		-60		dB
		with C <sub>i</sub> =2uF, V(ripple)=200mV	f=1kHz		-62		dB
CMRR	Common mode rejection ratio	V <sub>DD</sub> =3.6V, V <sub>ic</sub> =1Vpp	f=217Hz		-62		dB
twu	Device wake up time	V <sub>DD</sub> =5V, C <sub>BYP</sub> =0.1uF			150		ms
R <sub>PD</sub>	/SHDN pin internal pull-down resistor	V <sub>DD</sub> =2.5V to 5.5V			290		kΩ





## **Typical Performance Characteristics**

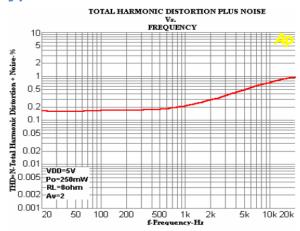
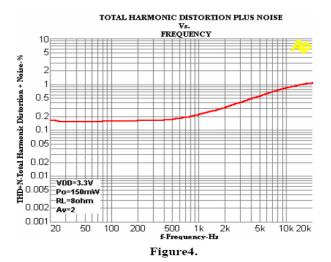
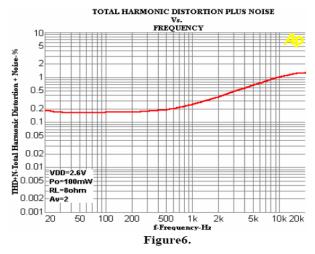


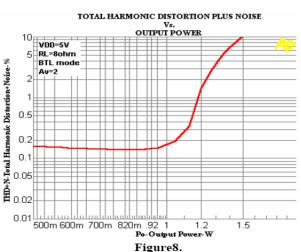
Figure3.



TOTAL HARMONIC DISTORTION PLUS NOISE Vs. FREQUENCY 10<sub>E</sub> HD+N-Total Harmonic Distortion + Noise-% 0.5 0.2 0.1 0.05 0.02 0.01 0.005 Po=100mW 0.002 500 1k f-Frequency-Hz 50 100 200 10k 20k Figure7.

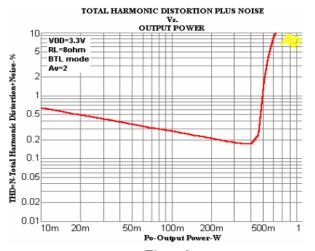














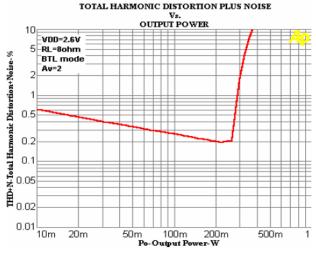
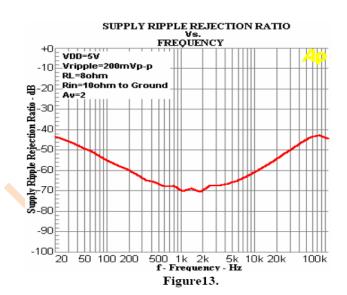


Figure11.



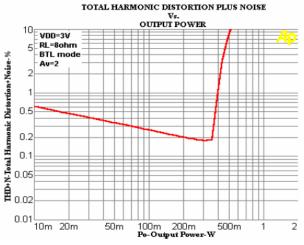


Figure 10.

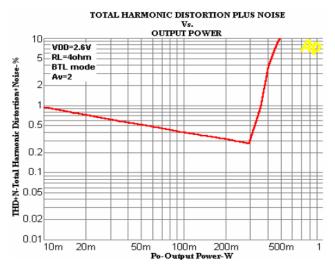
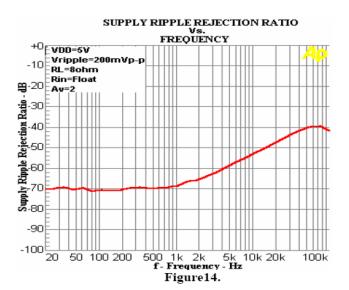
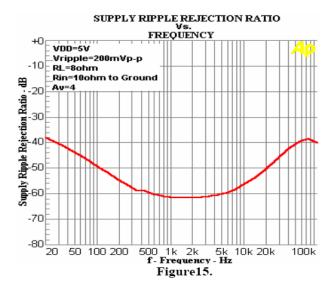


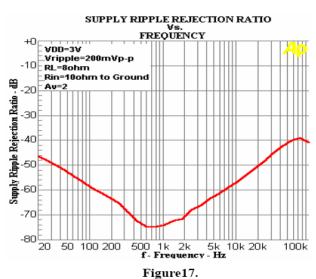
Figure 12.

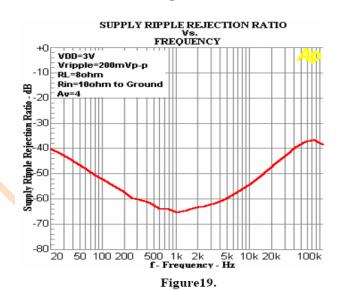


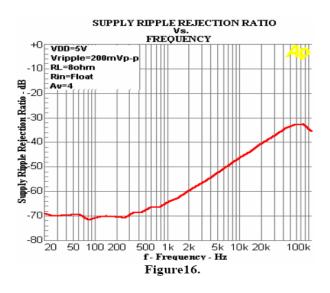


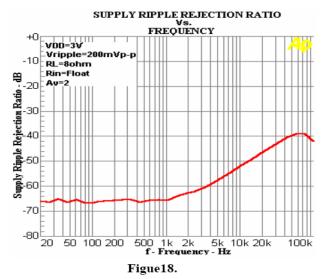


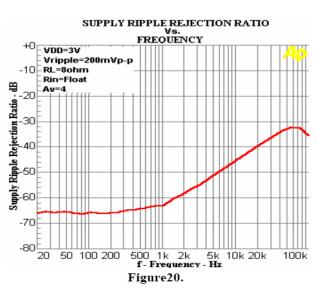






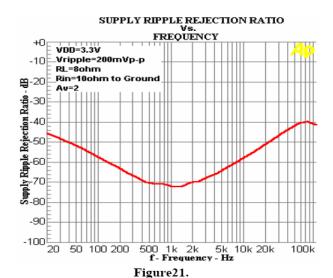












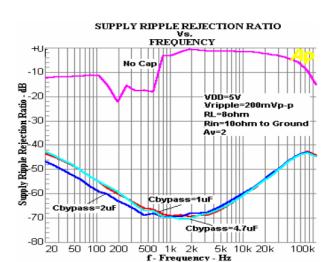
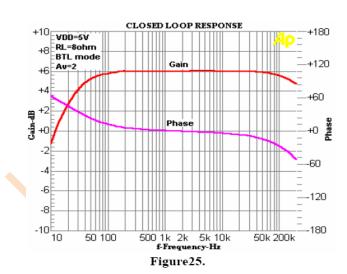


Figure 23.



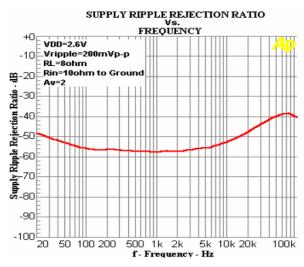
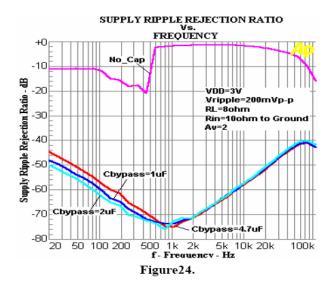


Figure 22.



CLOSED LOOP RESPONSE +15 +180 +12.5 +120 +10 +7.5 +60 +5 **Cair 4B** Phase +0 -60 VDD=5V -5 RL=8ohm BTL mode -120 -10<sup>t</sup> -180 10 50 100 500 1k 2k 5k 10k 50k 200k f-Frequency-Hz Figure 26.



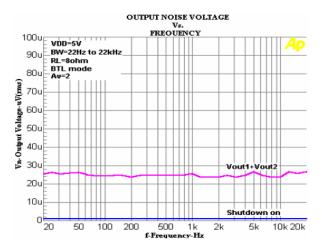


Figure 27.

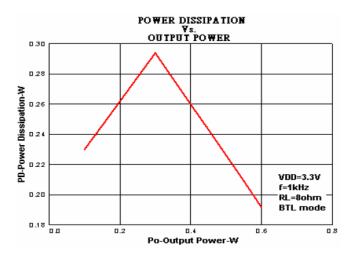
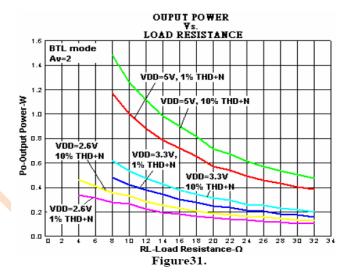


Figure 29.



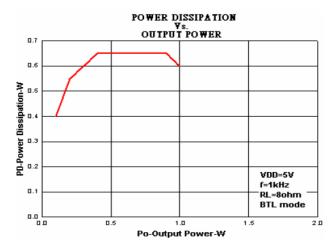


Figure 28.



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# **Functional Description**

### **General Description**

As shown in Figure 2, the LPA4890 has two operational amplifiers internally, allowing for a few different amplifier configurations. The first amplifier's gain is externally configurable, while the second amplifier is internally fixed in a unity-gain, inverting configuration. The close loop gain of the first amplifier is set by selecting the ratio of  $R_{\text{F}}$  to  $R_{\text{IN}}$  while the second amplifier's gain is fixed by the two internal  $20k\Omega$  resistors. Figure 2 shows the output of amplifier one serves as the input to amplifier two, which results in both amplifiers producing signals identical in magnitude, but out of phase by 180°. Consequently, the differential gain  $A_{\text{VD}}$  for the IC is

$$A_{VD} = 2 \times \frac{R_F}{R_{IN}}$$

By driving the load differentially through outputs VO1 and VO2, an amplifier configuration commonly referred to as "BTL mode" is established. The BTL mode operation is different from the classical single-ended amplifier configuration where one side of the load is connected to ground.

A BTL amplifier design has a few distinct advantages over the single-ended configuration, as it provides differential drive to the load, thus doubling output swing for a specified supply voltage. Four times the output power is possible as compared to a single-ended amplifier under the same conditions. This increase in attainable output power assumes that the amplifier is not current limited or clipped.

A BTL configuration, such as the one used in the LPA4890, also creates a second advantage over single-ended amplifiers. Since the differential outputs, VO1 and VO2, are biased at half-supply, no net DC voltage exists across the load. This eliminates the need for an output coupling capacitor which is required in a single supply, single-ended amplifier configuration. Without an output coupling capacitor, the half-supply bias

across the load would result in both increased internal IC power dissipation and also possible loudspeaker damage.

### **Power Supply Bypassing**

As with any amplifier, proper supply bypassing is critical for low noise performance and high-power supply rejection. The capacitor location on both the bypass and power supply pins should be as close to the device as possible. Typical applications employ a 5V regulator with 10 µF tantalum or electrolytic capacitor and a ceramic bypass capacitor that aid in supply stability. This does not eliminate the need for bypassing the supply nodes of the LPA4890. The selection of a bypass capacitor, especially C<sub>BYP</sub>, is dependent upon PSRR requirements, click and pop performance, system cost, and size constraints.

#### **Shutdown Mode**

In order to reduce power consumption while not in use, the LPA4890 contains a /SHDN pin to externally turn off the amplifier. This shutdown feature turns the amplifier off when a logic low is placed on the /SHDN pin. The LPA4890 has an internal pull-down resistor (R<sub>PD</sub>) on the /SHDN pin, if /SHDN pin is floating, the device will shut down. By switching the /SHDN pin to logic low or pull it down to ground, the LPA4890 supply current draw will be minimized to shutdown current (I<sub>SHDN</sub>), which is 0.1uA in typical.

#### **Thermal Shutdown**

The LPA4890 has thermal shutdown protection to fully protect the device from internally or externally generated excessive temperatures. The thermal shutdown circuit is activated when the die temperature exceeds a safe temperature (typ 170°C) and the switch is turned off. The switch automatically turns on again if the temperature drops below the threshold temperature.





## **Application Information**

### **Capacitor Consideration**

The external capacitor on VDD is recommended in application,  $1\mu F$  for  $C_{VDD}$  at least. Closer placement of the capacitors to the device is better for stability.

#### **Shutdown Mode**

The device contains a /SHDN pin to externally turn off the amplifier. In many applications, a microcontroller or microprocessor output is used to control the shutdown circuitry to provide a quick, smooth transition into shutdown. Another solution is to use a single-pole, single-throw switch S1 in conjunction with an external pull-up resistor, refer to Figure 2. When the switch connects the /SHDN pin to ground, the amplifier will be disabled. If the switch connects the /SHDN pin to VDD, the device will be enabled. The external pull-up resistor and internal pull-down resistor ( $R_{PD}$ ) will form a resistor divider, so choose a proper external pull-up resistor is important. Selecting a  $10 \text{K}\Omega$  external pull-up resistor or pulling the /SHDN pin to VDD directly is recommended. This scheme ensures the /SHDN pin in a certain state, thus preventing the unwanted state changes.

### **Power Dissipation**

Power dissipation is a major concern when designing a successful amplifier, whether the amplifier is bridged or single-ended. A direct consequence of the increased power delivered to the load by a bridge amplifier is an increase in internal power dissipation. Since the LPA4890 has two operational amplifiers in one package, the maximum internal power dissipation is 4 times of a single-ended amplifier. The maximum power dissipation for a given application can be derived from below Equation.

$$P_{\rm D} = 4 * V_{DD}^2 / (2\pi^2 R_L)$$

P<sub>D</sub>: Power Dissipation (W)

V<sub>DD</sub>: Input voltage (V)

 $R_L$ : Speaker resistance ( $\Omega$ )

$$T_{J} = P_{D} \times \theta_{JA} + T_{A}$$

T<sub>J</sub>: Junction temperature (°C)

θ<sub>JA</sub>: Package thermal resistance (°C /W) (Note 3)

T<sub>A</sub>: Ambient temperature (°C)

It is critical that the maximum junction temperature T<sub>JMAX</sub> of 150°C is not exceeded. T<sub>JMAX</sub> can be determined from the power derating curves by using P<sub>DMAX</sub> and the PC board foil area. By adding additional copper foil, the thermal resistance of the application can be reduced, resulting in higher P<sub>D</sub>. Additional copper foil can be added to any of the leads connected to the LPA4890. If T<sub>JMAX</sub> still exceeds 150°C, additional changes must be made. These changes can include reduced supply voltage, higher load impedance, or reduced ambient temperature.

\*Note 3: The calculation base on thermal resistance is only valid in Lab condition. The value of θ<sub>JA</sub> could change in customer PCB environment.

#### **PCB Layout Guidelines**

The PCB layout is critical to the optimal performance of an audio amplifier, some general mixed signal layout recommendations for LPA4890 as below:

- 1. Place the C<sub>VDD</sub> capacitor as close as possible to the device with short, wide traces to the VDD and GND pins.
- 2. The power ground should be connected to the analog ground through a single point. It is further recommended to put analog and power traces over the corresponding analog and power ground traces to minimize noise coupling.
- 3. The PCB traces that connect the output pins to the load and the supply pins to the power supply should be as wide as possible to minimize trace resistance, this is helpful to maintain the highest output voltage swing and





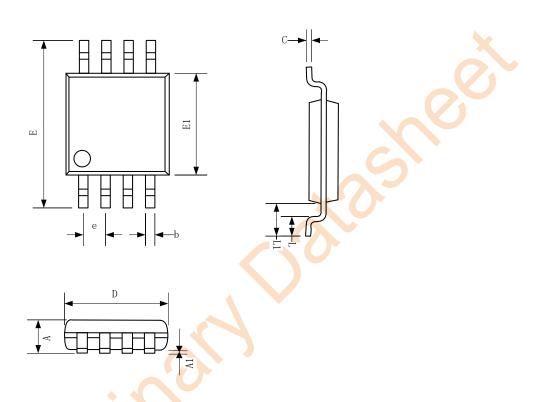
corresponding peak output power.

- 4. Avoid ground loops or running digital and analog traces parallel to each other (side-by-side) on the same PCB layer. Running digital and analog traces at 90 degrees to each other from the top to the bottom side as much as possible will minimize capacitive noise coupling and cross talk.
- 5. Ensure enough copper area is used for heat sinking to keep the junction temperature below 150°C.



# **Package Information**

# MSOP-8



SYMBOL	Dimensions In Millimeters			
STIVIDUL	MIN	NOM	MAX	
Α	-		1.10	
A1	0.00	-	0.15	
b	0.22	0.30	0.38	
С	0.17REF			
D	2.90	3.00	3.10	
E	4.70	4.90	5.10	
E1	2.90	3.00	3.10	
е	0.65BSC			
L	0.40	0.60	0.80	
L1	0.95 REF			